

Analogue Realization of Fractional-Order Healthy and Cancerous Lung Cell Models for Electrical Impedance Spectroscopy

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Abstract—This work proposes an integrated-circuit architecture emulating healthy and cancerous lung cell behaviors, approximated by Cole models and is suitable for calibration and phantom experimental testing of electrical bioimpedance circuits and systems. The architecture is based on fractional-order elements implemented with both active and passive components, offering an accurate transfer function behavior between 10kHz and 1MHz. The high-level architecture includes an analog all-pass filter coupled with a current conveyor. Performance and accuracy of the proposed architecture is confirmed via Monte-Carlo simulation. The proposed circuitry has been designed in TSMC 90nm CMOS process and simulated using the Cadence IC suite.

Index Terms—Fractional-order models, healthy lung cell, cancerous lung cell, current feedback operational amplifier

I. INTRODUCTION

The human tissues are characterized by specific electrical properties that reveal important information about their physical structure and function over space, time and frequency. Therefore, bioimpedance measurements, such as simple tetrapolar measurement, Electrical Impedance Tomography (EIT) and Electrical Impedance Spectroscopy (EIS) are recently becoming widespread methods that can assist the diagnostic process, avoiding invasive or continuous ionizing radiation exhibition procedures. In bioimpedance measuring methods, a small amplitude alternating current that meets the patient safety standards is acted via an electrode pair, attached on the patient skin surface [1].

Since bioimpedance measuring approaches are still developing both in hardware and software terms, accurate simulations and calibration experiments that include the subject electrical equivalent are immensely desired. This requirement introduces some challenges, since the electrical behavior of most human tissues is very complex and non-linear over the frequency domain. They usually follow a Cole frequency dispersion sequence. Moreover, the measurements often need to be taken at a wide frequency range, since each frequency region reveals specific tissue properties that might not be detectable at other regions [2].

Hence, high accuracy and large spectrum models are developed in order to replicate the human tissues' cells behavior.

Due to their Cole nature, fractional-order models are essential to achieve the desired accuracy. Many mathematical fractional-order models have been established for biological applications. However, there is a noticeable lack of circuit modelling realization, which is necessary for the simulation of bioimpedance measuring system and the calibration with prior experiments using phantom circuitries. This absence is presented due to the fact that Cole models are impossible to be implemented using linear RC circuit equivalents.

EIT and EIS are widely utilized at the thoracic imaging, for the examination of the lung's structural and functional characteristics. Due to the lung's dynamic behavior over the time domain, thoracic tomographic images are usually obtained at frequencies between 10kHz and 1MHz [1]. Moreover, any malignant tissue regions are more detectable at higher frequencies. Therefore, fractional-order models are needed to be applied at this frequency span.

In this work, an analogue implementation of fractional-order models, emulating normal and cancerous lung tissue cells from 10kHz-1MHz, is presented. Since fractional-order models offer more degrees of freedom, in comparison with integer-order, they are more accurate and sophisticated for real-world system description [3], [4]. The implementation of the models is based on analog filters. In order to transform the voltage (filter output) to impedance, we add a voltage to current (V/I) converter. The whole architecture describes the behavior of a fractional-order element.

The remainder of this paper is organized as follows. In Section II, a brief presentation of the mathematical fractional-order model for the healthy and the cancerous lung tissue is written. In Section III, the analogue implementation is described and in Section IV, simulations performed in Cadence IC suite and the corresponding results are demonstrated and discussed. Finally, in Section V, the conclusion is written.

II. LUNG COLE MODEL

Many studies related with the lungs' electrical properties over various frequency ranges are found in the literature. Extensive studies for Cole parameter fitting at a large variety of human tissues are found in [5], [6]. General Cole modelling

TABLE I: Normal and cancerous lung cell Cole parameters [11]

Parameter	R_∞ (Ω)	R_o (Ω)	a	τ ($\cdot 10^{-7}$ s)	C ($\mu F/s^{1-a}$)
Normal	153 ± 69	1280 ± 421	0.66 ± 0.06	5.70 ± 3.71	0.067
Cancerous	76 ± 80	558 ± 143	0.64 ± 0.08	6.82 ± 0.12	0.235

studies targetted in EIS are also presented in [7] and [8]. Specific work on the lung tissue cell electrical behavior from 1 MHz to 20GHz is found in [9], while [10] searches an accurate correlation between the air content and the admittance. In this particular work, we implement the Cole model described in [11], since it covers the frequency spectrum of interest (10 kHz to 1MHz) for EIS applications, and separates the model behavior between benign and malignant lung tissue cells.

Each cell can be modelled using a parallel combination of a Constant Phase Element (CPE) and a resistor, in series with a shunt impedance as it is depicted in Fig. 1. The Cole equation that expresses the healthy lung cell [11] is given by:

$$z_b(\omega) = R_{\infty,b} + \frac{R_{0,b} - R_{\infty,b}}{1 + (j\omega)^{a_b} \cdot (R_{0,b} - R_{\infty,b})C_b} \quad (1)$$

while the equation that expresses the cancerous cell [11] is given by:

$$z_m(\omega) = R_{\infty,m} + \frac{R_{0,m} - R_{\infty,m}}{1 + (j\omega)^{a_m} \cdot (R_{0,m} - R_{\infty,m})C_m} \quad (2)$$

Where $R_{\infty,i}$ is the shunt Ohmic impedance, towards which the cell's impedance converges while frequency tends to infinity, $R_{o,i}$ is the low frequency resistor, a_i is the fractional CPE order, ω is the angular frequency and C_i the CPE's pseudo-capacitance. The characteristic time constant τ_i is expressed by:

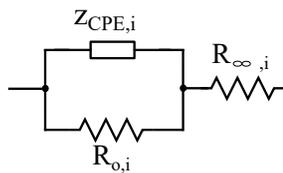
$$\tau_i = \sqrt[{}^i]{(R_{0,i} - R_{\infty,i})C_i} \quad (3)$$

The CPE's impedance is computed with the following expression:

$$z_{CPE} = \frac{1}{C_i(j\omega)^{a_i}} \quad (4)$$

It is noted that $i = b, m$, where b denotes a healthy lung cell and m a malignant one. The Cole parameter values for both cases are demonstrated in Table I, where the CPEs' values have been estimated from (3) using the mean parameter values.

It is obvious that there is a large variation at the parameters' values, mainly due to the lung's air content. A lung can be at any state between the full inhalation (inflated state) and the full exhalation (deflated state). Nevertheless, in this work we

Fig. 1: Cole model for a lung cell, where $i = b, m$.

are focusing on EIS applications for cancerous cell detection. Since between 10kHz and 1MHz the electrical conductivities and permittivities of a normal and a cancerous cell do not overlap [11], our analogue circuitry model is based on the mean parameter values.

III. MODEL IMPLEMENTATION

Implementation of the behavior of fractional-order elements is based on two types of approximation techniques due to the fact that they are not yet available for massive production. The first technique is RC-network approximation which provides easy implementation but it does not have the capability of electronic tuning on the element's characteristics (order or impedance) [12], [13].

The second one is integer order expression and the behavior of the element is approximated for example by means of the continue fraction expansion (CPE) method, round a center frequency $\omega_o = 1/\tau$. This is achieved via electronic parts such as operational transconductance amplifiers (OTAs) [14], current mirrors [15] and current feedback operation amplifiers (CFOAs) [16], [17]. A fractional-order model of lung is implemented in [14], using OTAs as active elements but the frequency range does not exceed 10Hz.

A. RC-network approximation

A typical network for approximating the behavior of fractional-order capacitor is depicted in Fig. 2. It consists of two correction factors R_p and C_p and m parallel resistors and capacitors branches. The generalized expression of the RC-network impedance is given by:

$$Z_{tot}(s) = sC_p + \frac{1}{R_p} + \sum_{\kappa=1}^m \frac{sC_\kappa}{sR_\kappa C_\kappa + 1} \quad (5)$$

Since the practical frequency of interest is between 10 kHz and 1MHz, the center frequency is $f_o = 100$ kHz and the appropriate order of the RC-network is $m = 3$ (in order to achieve at least 1.5° accuracy). Elements' values are summarized in Table II for both lung cells. The center frequency, the order m of the RC-network and the values of passive elements are derived using MATLAB code [12], [13].

B. CFE approximation

The behavior of fractional-order capacitor is described by a fractional-order differentiator connected with a double-output Current Conveyor (DO-CCII) which acts as Voltage to Current

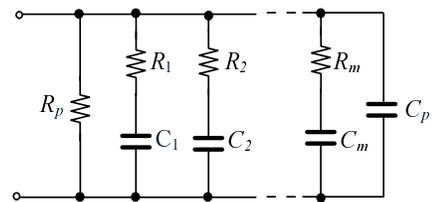


Fig. 2: Valsa-Vlach RC-network.

TABLE II: Passive element values for approximating the fractional-order capacitors of Fig. 2.

Normal Lung Element	Value	Cancerous Lung Element	Value
C_1	1.0 nF	C_1	2.90 nF
C_2	451.40 pF	C_2	1.25 nF
C_3	203.48 pF	C_3	535.77 pF
C_p	167.02 pF	C_p	404.42 pF
R_1	15.9 k Ω	R_1	5.5 k Ω
R_2	3.4 k Ω	R_2	1.2 k Ω
R_3	720 Ω	R_3	274 Ω
R_p	58.7 k Ω	R_p	19.1 k Ω

(V/I) converter. Fractional-order differentiator is approximated by an all-pass filter, which is implemented by utilizing Current Feedback Operational Amplifiers (CFOAs) as active elements.

The frequency range is set to be from 10kHz to 1MHz for our model and we choose 3rd-order CFE approximation round a center frequency $\omega_o = 1/\tau$ [14]–[17]. The expression of the 3rd-order CFE approximation is described by:

$$(\tau s)^\alpha \approx \frac{a_3 s^3 + a_2 s^2 + a_1 s + a_o}{b_3 s^3 + b_2 s^2 + b_1 s + b_o} \quad (6)$$

where:

$$\begin{aligned} \alpha_3 &= b_o = \alpha^3 + 6\alpha^2 + 11\alpha + 6, \\ \alpha_2 &= b_1 = -3\alpha^3 - 6\alpha^2 + 27\alpha + 54, \\ \alpha_1 &= b_2 = 3\alpha^3 - 6\alpha^2 - 27\alpha + 54, \\ \alpha_o &= b_3 = -\alpha^3 + 6\alpha^2 - 11\alpha + 6 \end{aligned}$$

and α is the order of the differentiator (all pass filter) [14]–[17]. The complete design of proposed architecture is depicted in Fig. 3 and parameters are summarized in Table III. In Fig. 3 the resistors' and capacitors' values can be easily calculated using the corresponding parameters based on (6). The transfer function of the fractional-order differentiator, which is based on a 3rd-order all pass filter is given by:

$$H(s) = (\tau s)^\alpha = \frac{a_3 s^3 + a_2 s^2 + a_1 s + a_o}{b_3 s^3 + b_2 s^2 + b_1 s + b_o} \quad (7)$$

The impedance of the fractional-order capacitor is given by:

$$Z_{cap}(s) = \frac{R_{vi}}{H(s)} \quad (8)$$

where R_{vi} is the appropriate resistor of the V/I converter and $H(s)$ is the expression of the filter.

TABLE III: Parameter of the circuit in Fig. 3.

Parameter	Value	Parameter	Value
b_3	1	b_2	$\frac{1}{R_1 C_1}$
b_1	$\frac{b_2}{R_2 C_2}$	b_o	$\frac{b_1}{R_3 C_3}$
α_3	$\frac{R_6 R_8}{R_1 (R_6 + R_8)}$	α_2	$\frac{\frac{R_6 R_8}{R_1} + \frac{R_6 R_8}{R_2 (R_6 + R_8)}}{R_3 (R_6 + R_8)}$
α_o	$\frac{b_1 R_8}{R_7 (R_6 + R_8)}$	α_1	$\frac{R_6 R_8 \alpha_o + \frac{R_6 R_8}{R_2 (R_6 + R_8)}}{R_3 R_5 (R_6 + R_8)}$

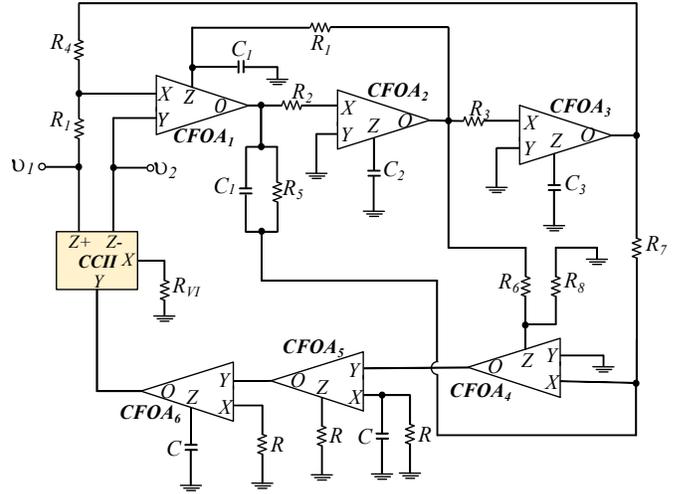


Fig. 3: Implementation of fractional-order capacitor emulator.

IV. SIMULATIONS AND RESULTS

The proposed lung cell has been designed in TSMC 90 nm CMOS process, using the Cadence IC design suite. The schematic of the corresponding CFOA is shown in Fig. 4 and the schematic of the appropriate V/I converter is depicted in Fig. 5, according to [17]. The power supply rails are set to $V_{DD} = -V_{SS} = 0.75$ V, $I_b = 200$ nA, $I_{bias} = 3$ nA and all transistors operate in the sub-threshold region. The dimensions of the MOS transistors of the CFOA and the DO-CCII are summarized in Table IV.

The obtained magnitude and phase responses in comparison with both RC-network and theoretically predicted ones confirm the behavior of the fractional-order healthy and cancerous lung cell model as shown in Fig. 6. The results of the proposed impedance models are in fine agreement with the theoretical ones and confirm the proper operation, performance and accuracy of the proposed topology.

The sensitivity behavior has been evaluated using the Monte-Carlo analysis tool for $N = 100$ runs for impedance and phase for health lung. The mean value of the impedance and phase is $I_{mean} = 1.06$ k Ω and $P_{mean} = -18.28^\circ$ and the standard deviation is $\sigma_I = 0.07$ k Ω and $\sigma_P = 1.36^\circ$

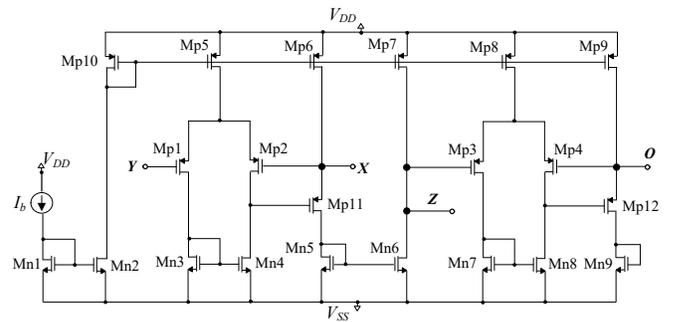


Fig. 4: Employed CFOA.

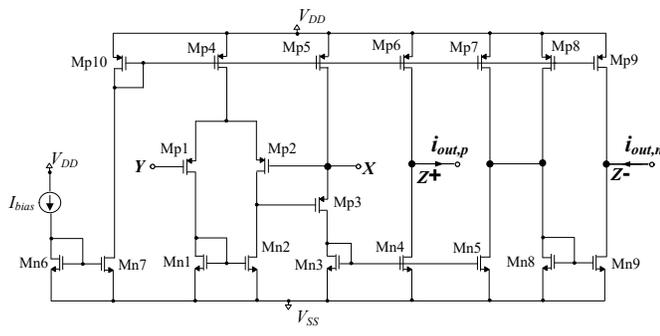


Fig. 5: Employed DO-CCII for the implementation of VI converter.

at $f_o = 100$ kHz, respectively. The Monte-Carlo analysis demonstrates the performance and accuracy of the proposed architecture.

TABLE IV: MOS Transistors Dimensions – CFOA & DO-CCII.

CFOA	W/L ($\mu\text{m}/\mu\text{m}$)	DO-CCII	W/L ($\mu\text{m}/\mu\text{m}$)
$M_{n1}-M_{n9}$	13/0.5	$M_{n1}-M_{n9}$	13/0.5
$M_{p5}-M_{p10}$	50/0.5	$M_{p4}-M_{p10}$	50/0.5
M_{p11}, M_{p12}	100/0.5	M_{p3}	100/0.5
$M_{p1}-M_{p4}$	20/0.4	M_{p1}, M_{p2}	20/0.4

V. CONCLUSION

In this paper, we implemented the healthy and cancerous lung cells (Cole model) in fractional order integrated form using CFOAs and CCII as active elements. Simulations showed very low impedance and phase errors for fractional-order model and also confirming the accuracy and the performance of the proposed architecture, in system level. The ICs implemented can be used in more complex circuitry setups

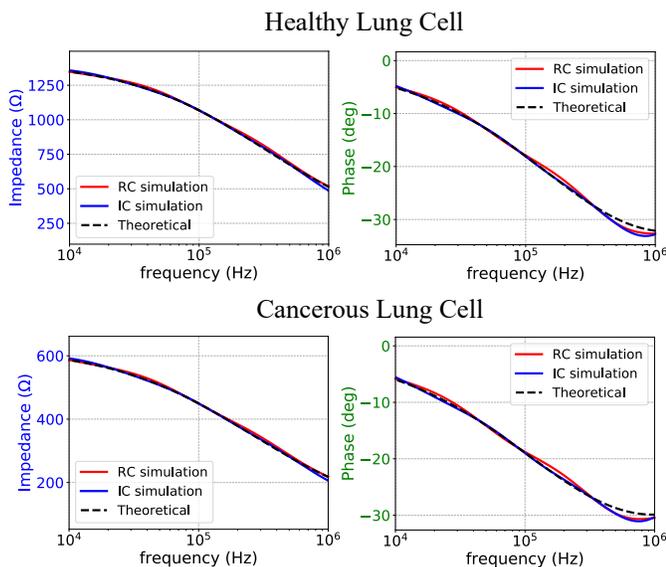


Fig. 6: Frequency responses of fractional-order lung cell model.

for calibration and phantom experimental testing, much more effectively than simple resistor or RC networks.

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